

## Chapter 10B

### FIR Digital Filter Design



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## Part B

### Computer-Aided Design of FIR Digital Filters



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### 1. Design of Equiripple Linear-Phase FIR Filters



- The linear-phase FIR filter obtained by minimizing the peak absolute value of

$$\varepsilon = \max_{\omega \in R} |E(\omega)|$$

which is usually called the **equiripple FIR filter**

- After  $\varepsilon$  is minimized, the weighted error function  $|E(\omega)|$  exhibits an equiripple behavior in the frequency range  $R$

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### 1. Design of Equiripple Linear-Phase FIR Filters



- The general form of frequency response of a causal linear-phase FIR filter of length  $2M+1$ :

$$H(e^{j\omega}) = e^{-jN\omega/2} e^{j\beta} \tilde{H}(\omega)$$

where the amplitude response  $\tilde{H}(\omega)$  is a real function of  $\omega$

- Weighted error function is given by

$$E(\omega) = W(\omega) [\tilde{H}(\omega) - D(\omega)]$$

where  $D(\omega)$  is the desired amplitude response and  $W(\omega)$  is a positive weighting function

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### 1. Design of Equiripple Linear-Phase FIR Filters



#### Parks-McClellan Algorithm

- Based on iteratively adjusting the coefficients of  $\tilde{H}(\omega)$  until the peak absolute value of  $E(\omega)$  is minimized
- If peak absolute value of  $E(\omega)$  in a band  $\omega_a \leq \omega \leq \omega_b$  is  $\varepsilon_0$ , then the absolute error satisfies

$$|\tilde{H}(\omega) - D(\omega)| \leq \frac{\varepsilon_0}{|W(\omega)|}, \quad \omega_a \leq \omega \leq \omega_b$$

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### 1. Design of Equiripple Linear-Phase FIR Filters



- For filter design,

$$D(\omega) = \begin{cases} 1, & \text{in the passband} \\ 0, & \text{in the stopband} \end{cases}$$

- $\tilde{H}(\omega)$  is required to satisfy the above desired response with a ripple of  $\pm\delta_p$  in the passband and a ripple of  $\delta_s$  in the stopband

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## 1. Design of Equiripple Linear-Phase FIR Filters



- Thus, weighting function can be chosen either as

$$W(\omega) = \begin{cases} 1, & \text{in the passband} \\ \delta_p / \delta_s, & \text{in the stopband} \end{cases}$$

or

$$W(\omega) = \begin{cases} \delta_s / \delta_p, & \text{in the passband} \\ 1, & \text{in the stopband} \end{cases}$$