

Chapter 4B

Digital Processing of Continuous-Time Signals



Part B

Design of Analog Filters



Design of Analog Filters

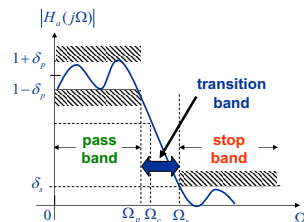


- ◆ Analog Lowpass Filter Specifications
- ◆ Butterworth Approximation
- ◆ Design of Other Types of Analog Filters

1. Analog Lowpass Filter Specifications



- Typical magnitude response of an analog lowpass filter may be given as indicated below



1. Analog Lowpass Filter Specifications



- In the **passband**, defined by $0 \leq \Omega \leq \Omega_p$, we require $1 - \delta_p \leq |H_a(j\Omega)| \leq 1 + \delta_p$, $|\Omega| \leq \Omega_p$ i.e., $|H_a(j\Omega)|$ approximates unity within an error of $\pm\delta_p$
- In the **stopband**, defined by $\Omega_s \leq \Omega \leq \infty$, we require $|H_a(j\Omega)| \leq \delta_s$, $\Omega_s \leq \Omega \leq \infty$ i.e., $|H_a(j\Omega)|$ approximates zero within an error of δ_s

1. Analog Lowpass Filter Specifications



- Ω_p -- passband edge frequency
- Ω_s -- stopband edge frequency
- δ_p -- peak ripple value **in the passband**
- δ_s -- peak ripple value **in the stopband**

Peak passband ripple

$$\alpha_p = -20 \log_{10}(1 - \delta_p) \text{ dB}$$

Minimum stopband attenuation

$$\alpha_s = -20 \log_{10}(\delta_s) \text{ dB}$$

2. Butterworth Approximation

- The magnitude-square response of an N -th order analog lowpass **Butterworth Filter** is given by

$$|H_a(j\Omega)|^2 = \frac{1}{1 + (\Omega/\Omega_c)^{2N}}$$
- First $2N - 1$ derivatives of $|H_a(j\Omega)|^2$ at $\Omega = 0$ are equal to zero
- The Butterworth lowpass filter thus is said to have a **maximally-flat magnitude** at $\Omega = 0$

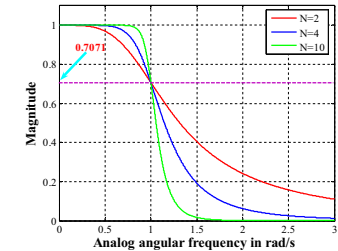
2. Butterworth Approximation

- Gain in dB is $G(\Omega) = 10 \log_{10} |H_a(j\Omega)|^2$ dB
- As $G(0) = 0$ dB and

$$G(\Omega_c) = 10 \log_{10}(0.5) = -3.0103 \approx -3 \text{ dB}$$
 Ω_c is called the **3-dB cutoff frequency**

2. Butterworth Approximation

- Typical magnitude responses with $\Omega_c = 1$



2. Butterworth Approximation

- Two parameters completely characterizing a Butterworth lowpass filter are Ω_c and N
- They are determined from the specified **bandedges** Ω_p and Ω_s , and **minimum passband magnitude** $|H_a(j\Omega_p)|$, and **maximum stopband magnitude** $|H_a(j\Omega_s)|$
- Ω_c and N are thus determined from

$$|H_a(j\Omega_p)|^2 = \frac{1}{1 + (\Omega_p/\Omega_c)^{2N}} \quad |H_a(j\Omega_s)|^2 = \frac{1}{1 + (\Omega_s/\Omega_c)^{2N}}$$

2. Butterworth Approximation

- According to the definition of the loss function

$$\alpha = 10 \log_{10} \frac{1}{|H(j\Omega)|^2}$$

- We know that $\frac{1}{|H(j\Omega)|^2} = 1 + \left(\frac{\Omega}{\Omega_c}\right)^{2N}$

$$1 + \left(\frac{\Omega_p}{\Omega_c}\right)^{2N} = 10^{\frac{\alpha_p}{10}} \quad 1 + \left(\frac{\Omega_s}{\Omega_c}\right)^{2N} = 10^{\frac{\alpha_s}{10}}$$

2. Butterworth Approximation

- Suppose

$$\lambda = \frac{\Omega_s}{\Omega_p} \quad k = \sqrt{\frac{10^{\frac{\alpha_p}{10}} - 1}{10^{\frac{\alpha_s}{10}} - 1}}$$

- Then

$$N = -\frac{\lg k}{\lg \lambda}$$

2. Butterworth Approximation

- Since order N must be an **integer**, value obtained is rounded up to the next highest integer
- This value of N is used next to determine by satisfying either the stopband edge or the passband edge specification exactly
- If the stopband edge specification is satisfied, then the passband edge specification is exceeded providing a safety margin

2. Butterworth Approximation

- Since $H_a(s)H_a(-s)$ evaluated at $s = j\Omega$ is simply equal to $|H_a(j\Omega)|^2$, it follows that

$$H_a(s)H_a(-s) = \frac{1}{1 + (-s^2/\Omega_c^2)^N}$$

- The poles of this expression occur on a circle of radius Ω_c at equally spaced points
- Because of the **stability** and **causality**, the transfer function itself will be specified by just the poles in the **negative real half-plane** of s

2. Butterworth Approximation

- Hence, the transfer function of an analog Butterworth lowpass filter has the form of

$$H_a(s) = \frac{C}{D_N(s)} = \frac{C}{s^N + \sum_{l=0}^{N-1} d_l s^l} = \frac{\Omega_c^N}{\prod_{l=1}^N (s - p_l)}$$

Where

$$p_l = \Omega_c e^{j[\pi(N+2l-1)/2N]}, \quad l = 1, 2, \dots, N$$

- Denominator $D_N(s)$ is known as the **Butterworth polynomial** of order N

2. Butterworth Approximation

Example –

- Determine the lowest order of a Butterworth lowpass filter with a 1 dB cutoff frequency at 1 kHz and a minimum attenuation of 40 dB at 5 kHz

$$10 \log_{10} \left[\frac{1}{1 + (2000\pi/\Omega_c)^{2N}} \right] = -1$$

$$10 \log_{10} \left[\frac{1}{1 + (10000\pi/\Omega_c)^{2N}} \right] = -40$$

$$N = \log_{10} \left\{ \frac{10^{0.1} - 1}{10^4 - 1} \right\} / \log_{10} \frac{1}{5} = 3.2811$$

3. Design of Other Types of Analog Filters

Highpass Filter

- **Step 1** Develop of **specifications** of a prototype analog lowpass filter $H_{LP}(s)$ from specifications of desired analog filter $H_D(s)$ using a **frequency (spectrum) transformation**.
- **Step 2** Design the prototype analog lowpass filter
- **Step 3** Determine the transfer function $H_D(s)$ of desired analog filter by applying the **inverse frequency transformation** to $H_{LP}(s)$

3. Design of Other Types of Analog Filters

- Let s denote the Laplace transform variable of prototype analog lowpass filter $H_{LP}(s)$ and \hat{s} denote the Laplace transform variable of desired analog filter $H_D(\hat{s})$
- The mapping from s -domain to \hat{s} -domain is given by the invertible transformation $s = F(\hat{s})$

Then

$$H_D(\hat{s}) = H_{LP}(s) \Big|_{s=F(\hat{s})}$$

$$H_{LP}(s) = H_D(\hat{s}) \Big|_{\hat{s}=F^{-1}(s)}$$

3. Design of Other Types of Analog Filters



- Spectral Transformation (Lowpass to Highpass)

$$s = \frac{\Omega_p \hat{\Omega}_p}{\hat{s}}$$

where Ω_p is the passband edge frequency of $H_{LP}(s)$ and $\hat{\Omega}_p$ is the passband edge frequency of $\hat{H}_{HP}(\hat{s})$

- On the imaginary axis the transformation is

$$\Omega = -\frac{\Omega_p \hat{\Omega}_p}{\hat{\Omega}}$$