

Ch7 LTI-Discrete-Time Systems in the Transform Domain

7.1 Let $G_L(z)$ and $G_H(z)$ represent ideal lowpass and highpass filters magnitude responses as sketched in Fig.1 a). Determine the transfer function $H_k(z) = Y_k(z) / X(z)$ of the discrete-time system of Fig.1 b), $k = 0, 1, 2, 3$, and sketch their magnitude response.

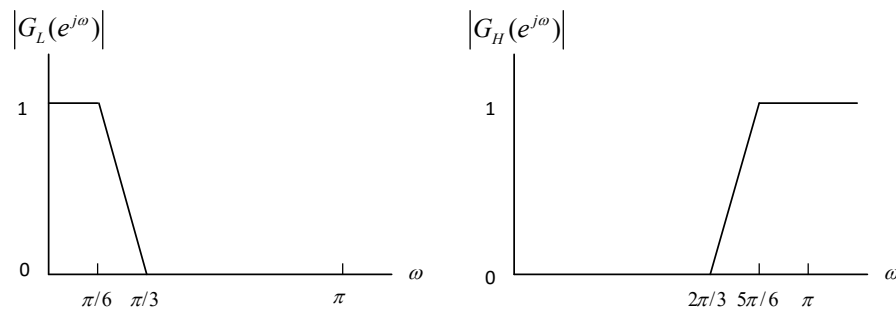


Fig. 1 a)

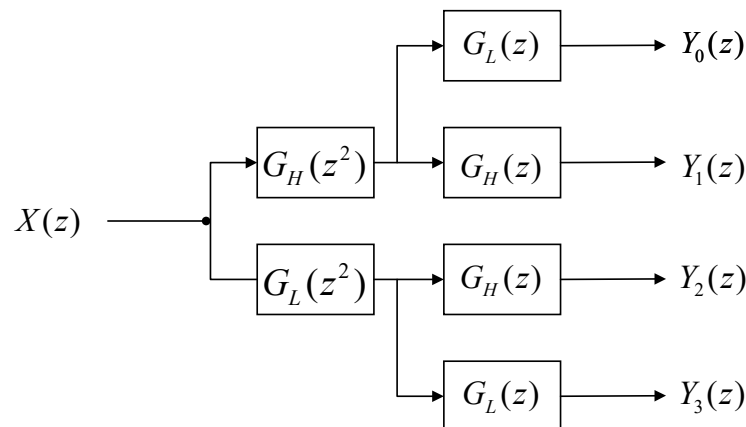


Fig. 1 b)

7.2 Let $H_{LP}(z)$ denote the transfer function of a real coefficient lowpass filter with a passband edge at ω_p , stopband edge at ω_s , passband ripple δ_p , and stopband ripple δ_s . Sketch the magnitude response of $G(z) = H_{LP}(-z)$, for $-\pi \leq \omega \leq \pi$. What type of filter is $G(z)$? Determine its frequency response $g[n]$ in terms of the impulse response $h_{LP}[n]$ of $H_{LP}(z)$. Determine the band edges and ripples of $G(z)$ in terms of the band edges and ripples of $H_{LP}(z)$.

7.3 Show that the structure shown in Fig.2 implements the filter $G(z)$ of Problem 7.2.

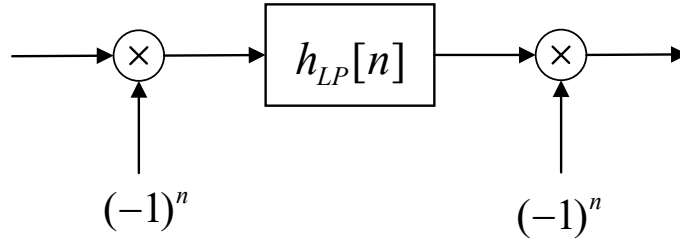


Fig. 2

- (Optional) Let $H_{LP}(z)$ denote the transfer function of an ideal real-coefficient lowpass filter having a cutoff frequency of ω_p , with $\omega_p < \pi/2$. Consider the complex coefficient transfer function $H_{LP}(e^{j\omega_0} z)$, where $\omega_p < \omega_0 < \pi - \omega_p$.
 - Sketch its magnitude response for $-\pi < \omega < \pi$. What type of filter does it represent?
 - Now consider the transfer function $G(z) = H_{LP}(e^{j\omega_0} z) + H_{LP}(e^{-j\omega_0} z)$. Sketch its magnitude response for $-\pi < \omega < \pi$.
 - Show that $G(z)$ is a real-coefficient bandpass filter with a passband centered at ω_0 .
 - Determine the width of its passband in terms of ω_p and its impulse response $g[n]$ in terms of the impulse response $h_{LP}[n]$ of the parent lowpass filter.
- (Optional) Show that the structure shown in Fig.3 implements the bandpass filter of optional problem above.

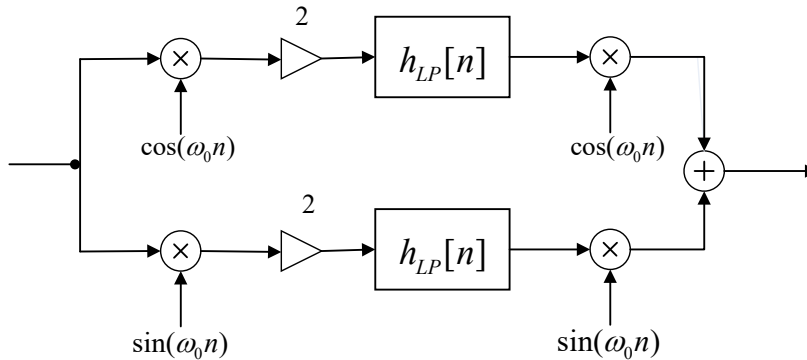


Fig. 3

7.4 Let $H(z)$ be a lowpass filter with unity passband magnitude, a passband edge at ω_p , and a stopband edge at ω_s , as shown in Fig. 4.

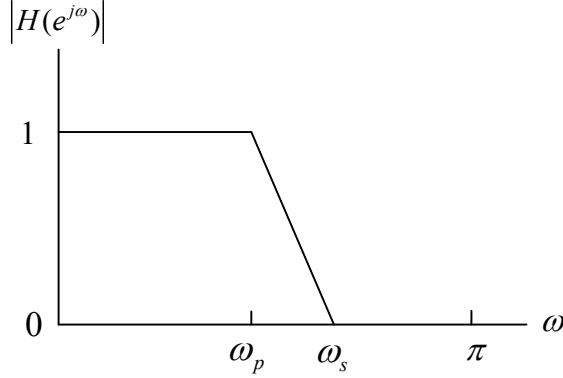


Fig. 4

- a) Sketch the magnitude response of the digital $G_1(z) = H(z^M)F_1(z)$, where $F_1(z)$ is a lowpass filter with unity passband magnitude, a passband edge at ω_p / M , and a stopband edge at $(2\pi - \omega_s) / M$. What are the band edges of $G_1(z)$?
- b) Sketch the magnitude response of the digital $G_2(z) = H(z^M)F_2(z)$, where $F_2(z)$ is a bandpass filter with unity passband magnitude, and with passband edges at $(2\pi - \omega_p) / M$ and $(2\pi + \omega_p) / M$, and stopband edges at $(2\pi - \omega_s) / M$ and $(2\pi + \omega_s) / M$, respectively. What are the band edges of $G_2(z)$?

7.5 Show analytically that the following causal FIR or IIR transfer functions are BR functions:

- a) $H_1(z) = \frac{1}{5}(1 + 4z^{-1})$
- b) $H_2(z) = \frac{(1 + \alpha z^{-1})(1 - \beta z^{-1})}{(1 + \alpha)(1 + \beta)}, \quad \alpha > 0, \quad \beta > 0$
- c) $H_3(z) = \frac{3.9 + 3.9z^{-1}}{3.4 + z^{-1}}$
- d) $H_4(z) = \frac{1.9 + 1.2z^{-1} + 1.9z^{-2}}{2 + 1.2z^{-1} + 1.8z^{-2}}$

7.6 Consider the first-order causal and stable allpass transfer function given by

$$A_1(z) = \frac{1 - d_1^* z}{z - d_1}$$

$$\text{Determine the expression for } 1 - |A_1(z)|^2 \begin{cases} < 0, & \text{for } |z|^2 < 1, \\ = 0, & \text{for } |z|^2 = 1, \\ > 0, & \text{for } |z|^2 > 1. \end{cases}$$

Now, using the above approach, show the Property 2 given by:

$$|A_M(z)| \begin{cases} < 1, & \text{for } |z| > 1, \\ = 1, & \text{for } |z| = 1, \\ > 1, & \text{for } |z| < 1. \end{cases}$$

holds for any arbitrary causal stable allpass transfer function.

7.7 Consider the transfer function given by

$$H(z) = \frac{1}{M} \sum_{k=0}^{M-1} \mathcal{A}_k(z)$$

where $\{\mathcal{A}_k(z)\}$ are stable real-coefficient allpass function. Show that $H(z)$ is a BR function.

7.8 Consider the cascade of two causal LTI discrete-time systems $h_1[n] = \alpha\delta[n] + \beta\delta[n-1]$

and $h_2[n] = \gamma^n \mu[n]$, $\beta < 1$. Determine the frequency response $H(e^{j\omega})$ of the overall system. For what value of α , β and γ will $|H(e^{j\omega})| = K$, where K is a **real constant**?

7.9 A noncausal LTI FIR discrete-time system is characterized by an impulse response

$$h[n] = a_1\delta[n-2] - a_2\delta[n-1] - a_3\delta[n] + a_4\delta[n+1] - a_5\delta[n+2]$$

For what values of the impulse response samples will its frequency response $H(e^{j\omega})$ have a **zero phase**.

● (Optional) A causal FIR discrete-time system is characterized by the difference function

$$y[n] = a_1x[n+k+1] + a_2x[n+k] + a_2x[n+k-1] + a_1x[n+k-2]$$

where $y[n]$ and $x[n]$ denote, respectively, the output and the input sequences. Determine

the expression for its frequency response $H(e^{j\omega})$. For what value of the constant k will the

system have a frequency response $H(e^{j\omega})$ that is a **real function of ω** ?

- 7.10 Let a causal LTI discrete-time system be characterized by a real impulse response $h[n]$ with a DTFT $H(e^{j\omega})$. Consider the system of Fig.5, where $x[n]$ is a finite-length sequence. Determine the frequency response of the overall system $G(e^{j\omega})$ in terms of $H(e^{j\omega})$ and show that it has a zero-phase response.

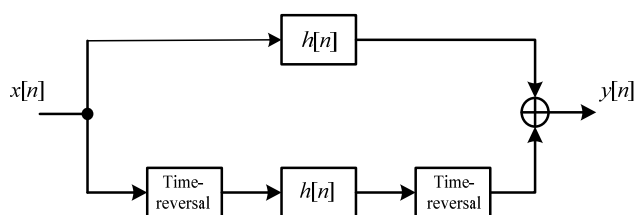


Fig. 5

- 7.11 Given causal IIR transfer functions:

$$H_1(z) = \frac{z^3 + 3z^2 + 2z + 7}{(3z - 4)(z^2 - 0.25)}$$

- Check the stability of the transfer functions.
- If it is not stable, find a stable transfer function with an identical magnitude function.
- Are there any other transfer functions having the same magnitude response as those shown above?

- 7.12 Determine all possible causal stable transfer function $H(z)$ with a squared-magnitude function given by

$$|H(e^{j\omega})|^2 = \frac{4(1.25 + \cos \omega)(1.36 - 1.2 \cos \omega)}{(1.36 + 1.2 \cos \omega)(1.64 + 1.6 \cos \omega)}$$

- (Optional) Consider the following FIR transfer functions:

- $H_1(z) = 1 - 0.52z^{-1} + 0.92z^{-2} + 0.87z^{-3} + 0.92z^{-4} - 0.52z^{-5} + z^{-6}$
- $H_2(z) = 0.6 + 0.652z^{-2} + 0.6928z^{-3} + 0.4032z^{-4} - 0.2z^{-5} - 0.12z^{-6}$
- $H_3(z) = 3.12 - 2.5z^{-1} + 0.6z^{-2} + 0.5z^{-3} + 0.06z^{-4} + z^{-5}$

Using the M-file *zplane*, determine the zero locations of each, and then answer the following questions:

- Is there any FIR filters having a linear-phase response?
- Is there any FIR filters having a minimum-phase response?

- ③ Is there any FIR filters having a maximum-phase response?

7.13 The transfer function of a type 2 linear-phase FIR filter is given by

$$H(z) = 2.5 + 0.5z^{-1} + 0.35z^{-2} + 5.47z^{-3} + 5.47z^{-4} + 0.35z^{-5} + 0.5z^{-6} + 2.5z^{-7}$$

- Determine the transfer function of $H_1(z)$ of a minimum-phase FIR filter having the same magnitude as that of $H(z)$.
- Determine the transfer function of $H_2(z)$ of a maximum-phase FIR filter having the same magnitude as that of $H(z)$.
- Determine the transfer function of $H_3(z)$ of a mixed-phase FIR filter having the same magnitude as that of $H(z)$.
- How many other length-8 FIR filters exist that have the same magnitude responses as that of $H(z)$.

7.14 Consider the following IIR transfer function of a causal digital filter:

$$\text{i. } H_1(z) = \frac{6 - 5z^{-1} + z^{-2}}{1 + 0.1z^{-1} - 0.56z^{-2}}$$

$$\text{ii. } H_2(z) = \frac{3 - 7z^{-1} + 2z^{-2}}{1 + 0.1z^{-1} - 0.56z^{-2}}$$

$$\text{iii. } H_3(z) = \frac{2 - 7z^{-1} + 3z^{-2}}{1 + 0.1z^{-1} - 0.56z^{-2}}$$

$$\text{iv. } H_4(z) = \frac{1 - 5z^{-1} + 6z^{-2}}{1 + 0.1z^{-1} - 0.56z^{-2}}$$

- Are the transfer function BIBO stable?
- What is the relation between their magnitude functions?
- What is the relation between their phase functions?

7.15 Is the transfer function

$$H(z) = \frac{(3z + 4)(z - 5)}{(z - 0.5)(z + 0.8)}$$

Minimum-phase? If it is not minimum-phase, the construct a minimum-phase transfer function $G(z)$ such that $|G(e^{j\omega})| = |H(e^{j\omega})|$. Determine their corresponding unit sample responses, $g[n]$ and $h[n]$, for $n = 0, 1, 2, 3, 4$. For what values of m is $\sum_{n=1}^m |g[n]|^2$ bigger than $\sum_{n=1}^m |h[n]|^2$.

- (Optional) Consider a causal minimum-phase transfer function $H_m(z)$ and a causal

non-minimum-phase transfer function $H(z)$, with identical magnitude response function $|H(e^{j\omega})| = s |H_m(e^{j\omega})|$. Let $h_m[n]$ and $h[n]$ denote the impulse responses of $H_m(z)$ and $H(z)$.

Prove:

a) $|h_m[0]| \geq s |h[0]|$

b) $\sum_{l=1}^n |h_m[l]|^2 \geq \sum_{l=1}^n |h[l]|^2$

(Property of causal minimum-phase transfer function)

7.16 A causal LTI FIR discrete-time system is characterized by an impulse response

$$h[n] = a_1 \delta[n] + a_2 \delta[n-1] + a_3 x[n-2] + a_4 x[n-3] + a_5 x[n-4] + a_6 x[n-5]$$

For what values of the impulse response samples will its frequency response $H(e^{j\omega})$ have a **constant group delay**?

7.17 Show that the amplitude response $\check{H}(\omega)$ of Type 1 and Type 3 linear-phase FIR transfer

functions is a periodic function of ω with a period 2π and the amplitude response $\check{H}(\omega)$ of Type 2 and Type 4 linear-phase FIR transfer functions is a periodic function of ω with a period 4π .

7.18 A type 2 real-coefficient FIR filter with a transfer function $H(z)$ has the following zeros: $z_1=1$, $z_2=-1$, $z_3=0.5$, $z_4=0.8+j$.

- Determine the locations of the remaining zeros of $H(z)$ having the lowest order.
- Determine the transfer function $H(z)$ of the filter.

- (Optional) We have shown that a real-coefficient FIR transfer function $H(z)$ with a symmetric impulse response has linear-phase response. As a result, the all-pole IIR transfer function $G(z) = 1/H(z)$ will also have a linear-phase response. What are the practical difficulties in implementing $G(z)$? Justify your answer.

7.19 The first 5 samples of the impulse response of an FIR filter $H(z)$ are given by $h[0] = 2$,

$$h[1] = 1.5, h[2] = -3.2, h[3] = -5.2, h[4] = 6.4.$$

- Determine the remaining impulse response samples of $H(z)$ of lowest order for each type of linear-phase filter;
- Using *zplane*, determine the zero locations for $H(z)$ for each type of linear-phase filter;

- c) Does $H(z)$ have a zero at $z = 1$ and/or $z = -1$;
- d) Do the zeros on the unit circle appear in complex conjugate pairs? Justify your answer.
- e) Do the zeros *not* on the unit circle appear in mirror-image symmetry? Justify your answer.

7.20 The phase response $\theta(\omega)$ of a linear-phase FIR filter is given by

$$\theta(\omega) = \begin{cases} -2\omega - \pi, & -\pi < \omega \leq 0, \\ \pi - 2\omega, & 0 < \omega \leq \pi. \end{cases}$$

What is the length of the impulse response of the filter and is it symmetric or anti-symmetric? What are the values of the magnitude response of the filter at dc and at $\omega = \pi$? Justify your answer.

7.21 Design a first-order lowpass IIR filter for each of the following normalized 3-dB cutoff frequencies:

- a) 0.42 rad/samples
- b) 0.65π

7.22 Design a first-order highpass IIR filter for each of the following normalized 3-dB cutoff frequencies:

- a) 0.5 rad/samples
- b) 0.45π