

1. Preliminary Considerations



- > Digital Filter Specifications
- > Selection of the Filter Type
- > Basic Approaches to Digital Filter Design
- **Estimation of the Filter Order**
- Scaling the Digital Filter

3

1.1 Digital Filter Specifications



Objective:

Determination of a realizable transfer function G(z) approximating a given frequency response specification is an important step in the development of a digital filter

- If an IIR filter is desired, G(z) should be a stable rational function
- Digital filter design is the process of deriving the transfer function G(z)

1.1 Digital Filter Specifications



- Usually, either the magnitude and/or the phase (delay) response is specified for the design of digital filter for most applications.
- In most practical applications, the problem of interest is the development of a realizable approximation to a given magnitude response specification.

1.1 Digital Filter Specifications



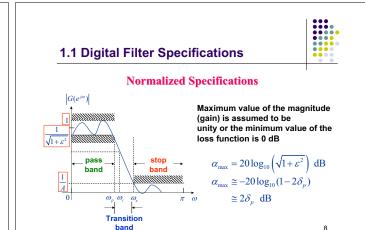
Digital Filter Design Steps

A choice between IIR and FIR digital filter has to be made

- ① Derivation of a realizable transfer function G(z)
- ② Realization of G(z) using a suitable filter structure

5

1.1 Digital Filter Specifications Digital Filter Specifications Passband: $|-\delta_p \le |G(e^{j\omega})|$ • Passband: $|-\delta_p \le |G(e^{j\omega})| \le 1 + \delta_p, \quad |\omega| \le \omega_p$ • Stopband: $|G(e^{j\omega})| \le \delta_s, \quad \omega_s \le \omega \le \pi$ • Peak Passband ripple: $\alpha_p = -20\log_{10}(1 - \delta_p) \text{ dB}$ • Minimum Stopband attenuation $\alpha_s = -20\log_{10}(\delta_s) \text{ dB}$ 7







• Frequency specifications are normalized using the sampling rate:

$$\omega_p = \frac{\Omega_p}{F_T} = \frac{2\pi F_p}{F_T} = 2\pi F_p T$$

$$\omega_s = \frac{\Omega_s}{F_T} = \frac{2\pi F_s}{F_T} = 2\pi F_s T$$

• $\omega = \pi$ corresponds to half the sampling rate, $F_{\pi}/2$

Q: What is the condition for non-overlapping?

9

1.2 Selection of the Filter Type



• FIR filters:

- ✓ Linear phase response
- ✓ Stability with quantized coefficients
- ⊠ Higher order required than using IIR filters

1.2 Selection of the Filter Type



11

• IIR filters:

- ✓ Better attenuation properties
- ✓ Closed form approximation formulas
- × Nonlinear phase response
- X Instability with finite wordlength computation
- ✓ Lower order

 $N_{\rm FIR}/N_{\rm IIR}$ is typically of the order of tens (or more)

1.3 Basic Approaches to Digital Filter Design



IIR Filter Design

An analog filter transfer function $H_a(s)$ is transformed into the desired digital filter transfer function G(z)

- Analog approximation techniques are highly advanced
- They usually yield closed-form solutions

1

1.3 Basic Approaches to Digital Filter Design



IIR Filter Design

- Extensive tables are available for analog filter design or the methods are easy to program
- Digital filters often replace (or simulate) analog filters

$$H_a(s) = \frac{P_a(s)}{D_a(s)} \Longrightarrow G(z) = \frac{P(z)}{D(z)}$$

13

1.3 Basic Approaches to Digital Filter Design



IIR Filter Design

• The basic idea behind the conversion of an analog prototype transfer function $H_a(s)$ to a digital filter transfer function G(z) is to apply a mapping from the *s*-domain to the *z*-domain so that the *essential properties* of the analog frequency response are preserved.

1.3 Basic Approaches to Digital Filter Design



IIR Filter Design

- Requirements for the transform are:
- > The imaginary axis $(j\Omega)$ of the s-plane is mapped onto the unit circle in the z-plane
- > Stable $H_a(s)$ must be transformed into a stable G(z)

15

1.3 Basic Approaches to Digital Filter Design



FIR Filter Design

- No analog prototype filters are available
- FIR filter design is based on a direct approximation of the specified magnitude response
- A linear phase response is usually required

1.3 Basic Approaches to Digital Filter Design



FIR Filter Design

• FIR transfer function:

$$H(z) = \sum_{n=0}^{N-1} h(n)z^{-n}$$

• The corresponding frequency response:

$$H(e^{j\omega}) = \sum_{n=0}^{N-1} h(n)e^{-j\omega n}$$

• Linear phase requirement:

$$h(n) = \pm h(N-1-n)$$

17

1.4 Estimation of the Filter Order



- IIR Design -- Filter order is solved from the approximation formulas
- FIR Design -- Several formulas proposed for estimating the minimum length of the impulse response

Kaiser: $N \approx \frac{-20 \log_{10} \left(\sqrt{\delta_p \delta_s} \right) - 1}{14.6 (\omega_s - \omega_n) / 2\pi}$



1.4 Estimation of the Filter Order

- N is inversely proportional to the normalized transition width and does not depend on the location of the transition band
- N depends also on the product of δ_n and δ_s



20

1.5 Scaling the Digital Filter

- G(z) has to be scaled in magnitude so that the maximum gain in the passband is unity
- Notice that the scaling coefficient *K* does not affect the shape of the magnitude response, i.e., it does not affect the locations of poles and zeros in the z-plane





• Lowpass filter: Unity gain at zero frequency $\omega = 0$ (or z=1)

$$KG(e^{j\omega})\Big|_{\omega=0} = KG(z)\Big|_{z=1} \longrightarrow K = 1/G(1)$$

• **Highpass filter:** Unity gain at $\omega = \pi$ (or z=-

1)
$$KG(e^{j\omega})\Big|_{\omega=\pi} = KG(z)\Big|_{z=-1} \longrightarrow K = 1/G(-1)$$

19